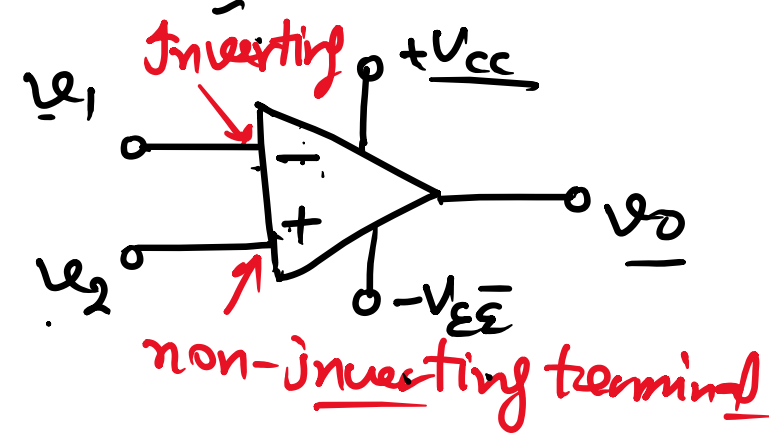
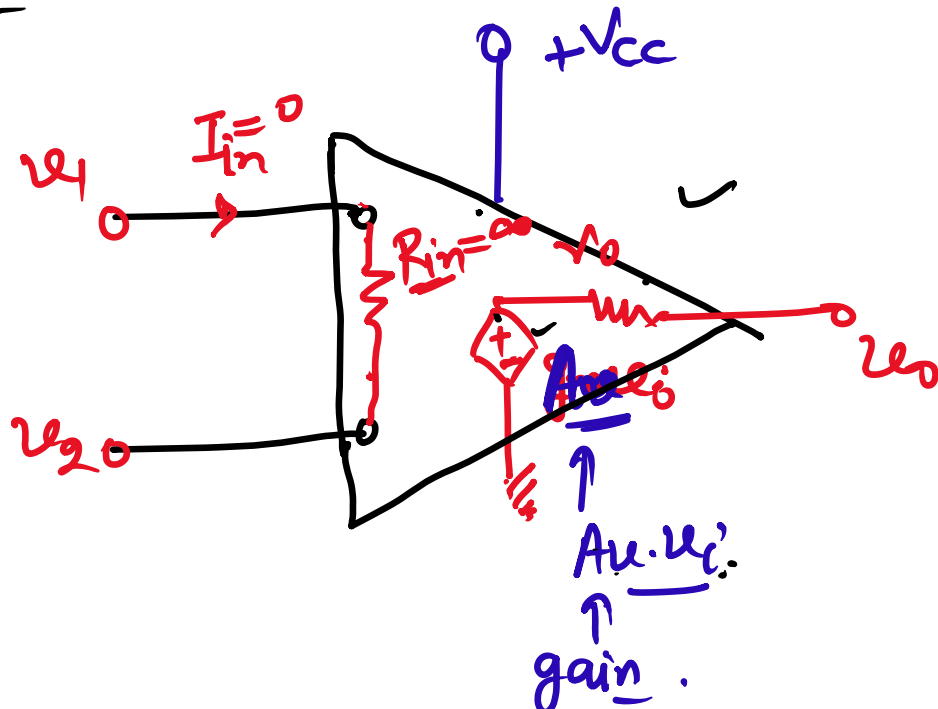


Unit-iv OP-Amp-

- ① Active device ✓
- ② made up of combinations of BJT or FET
- ③ Can operate with DC & AC.
- ④ Can be used in open loop & closed loop configurations.



Ideal OP-Amp-



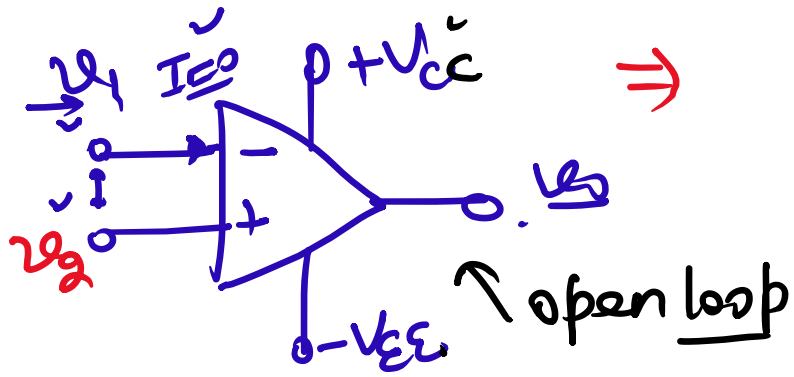
- ① $R_{in} = \infty \Rightarrow I_{in} = 0$
- ② Virtual short / Virtual ground.

The characteristics of ideal OP-Amp-

- ① $R_{in} = \infty \Rightarrow$ i/p resistance. -
- ② $R_o = 0 \Rightarrow$ o/p resistance
- ③ $A_v = \infty \Rightarrow$ Voltage Gain. $= \frac{V_o}{V_i}$

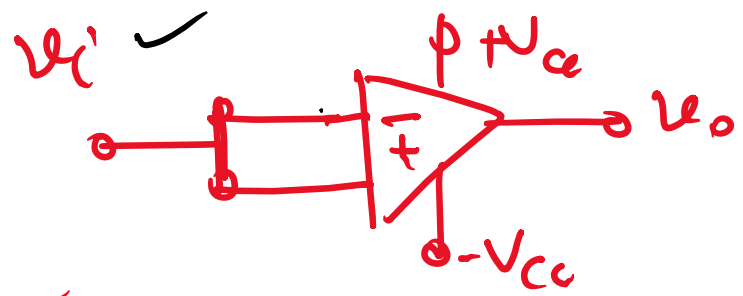
④ CMRR (Common mode rejection ratio) -

$$\underline{\text{CMRR}} = \frac{A_d}{A_c} \Rightarrow \frac{\text{differential mode gain } (A_d)}{\text{Common mode gain } (A_c)} \Rightarrow \underline{\underline{\infty}} \text{ (Ideal)}$$



$$A_v = \frac{v_o}{v_1 - v_2} = A_d$$

Offset
 $v_i = 0, \Rightarrow v_o \neq 0$
 ↑
 We have to apply opposite voltage or current so v/p become zero. -

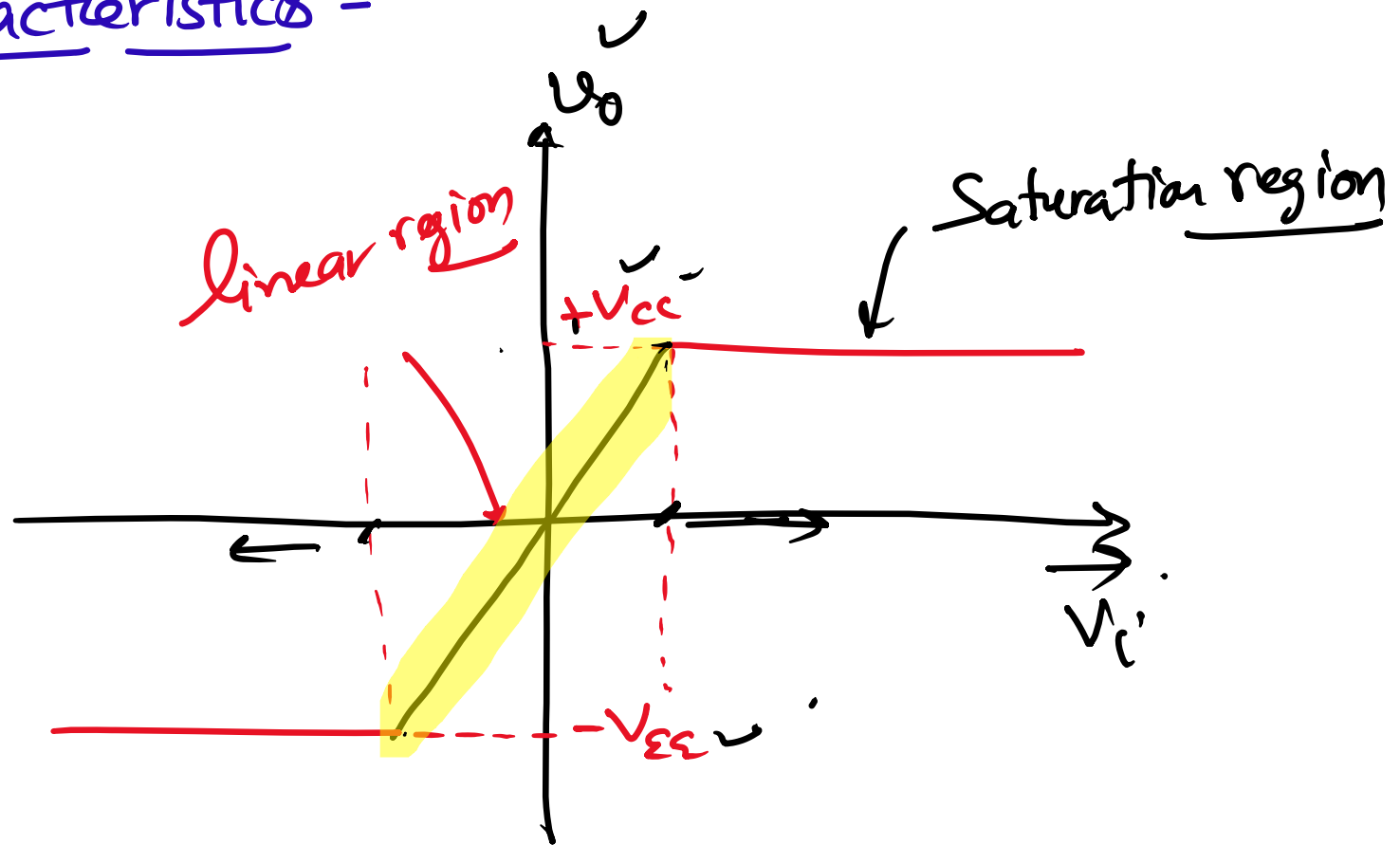


$$A_e = \frac{v_o}{v_i} \Rightarrow A_c$$

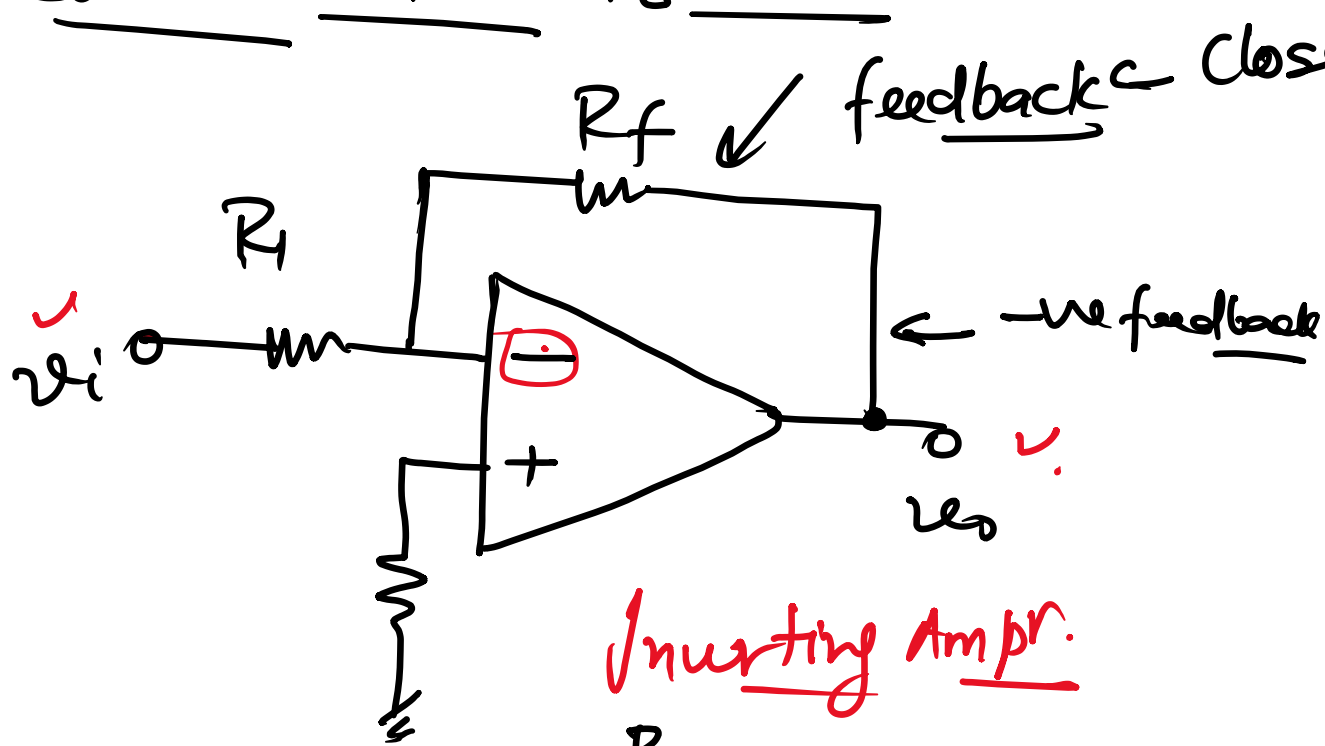
⑤ Slew rate = $S = \underline{\text{maximum of}} \frac{dV_o}{dt} \Rightarrow$ maximum rate of change of voltage with time
 $S \approx \underline{\infty}$ for ideal op-amp

Parameter	Ideal OP-Amp.	Practical OP-Amp.
① R_i	∞	Very high
② R_o	0	Very low
③ A_v	∞	Very high
④ CMRR	∞	Very high
⑤ SR	∞	Very high
⑥ <u>Input offset</u>	0	very small

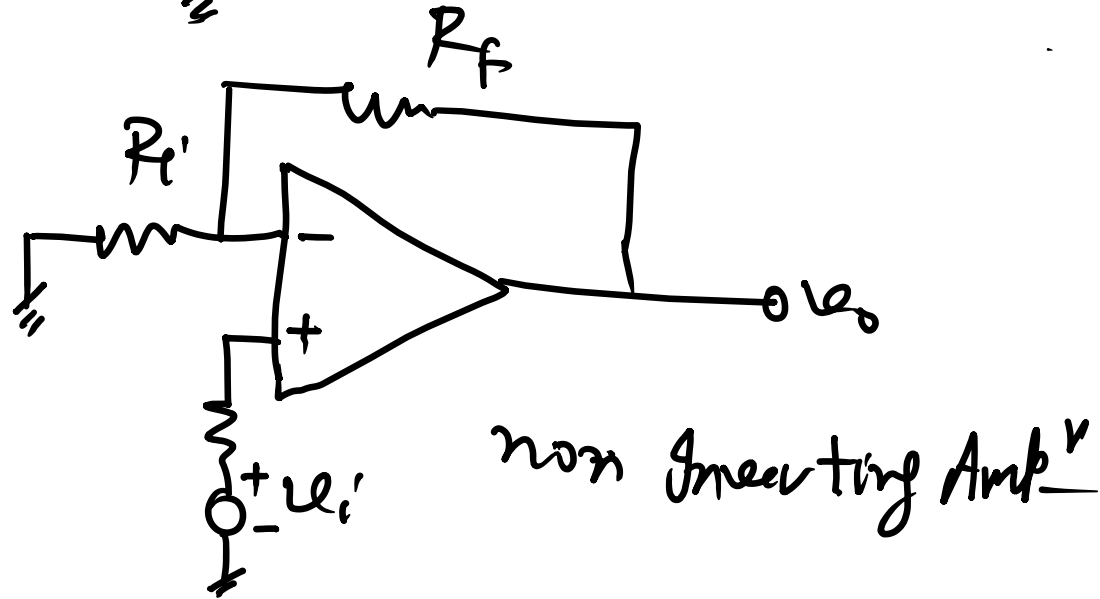
Characteristics -



⇒ Closed loop Configuration -



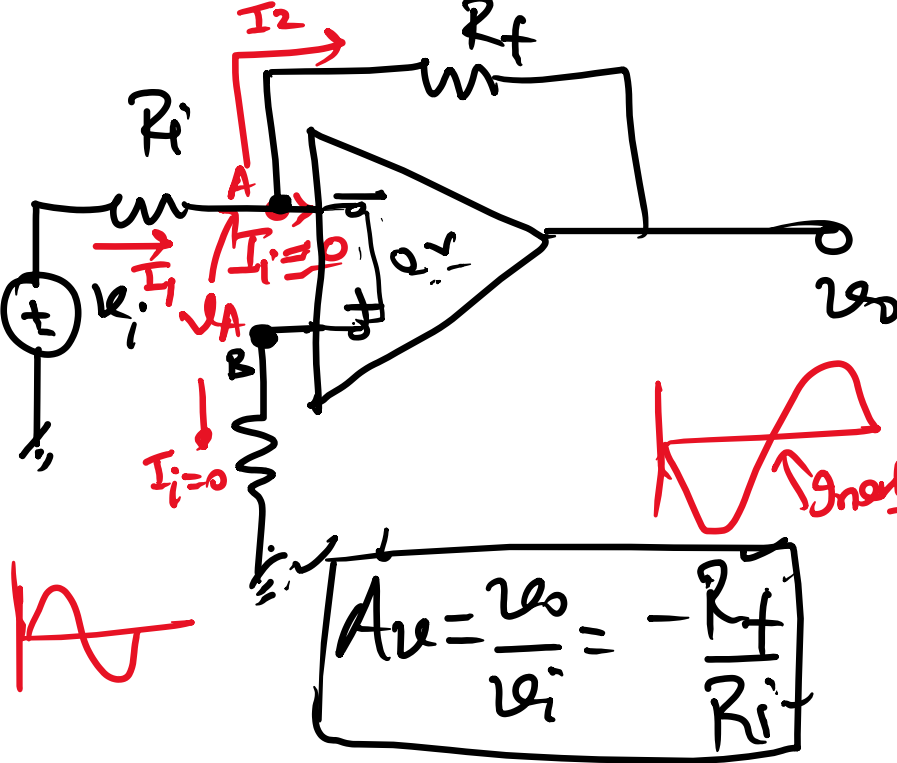
Inverting Amp^r



non Inverting Amp^r

feedback → +ve
-ve

-negative feedback → Amplifier
+ve feedback → Oscillator



$$A_v = \frac{v_o}{v_i} = -\frac{R_f}{R_i}$$

Apply KCL at node-d.

$$I_1 = I_2 - I_i \Rightarrow \underline{I_i = 0; R_{in} = \infty}$$

$$\Rightarrow I_1 = I_2$$

$$\frac{v_i - v_A}{R_i} = \frac{v_A - v_o}{R_f}$$

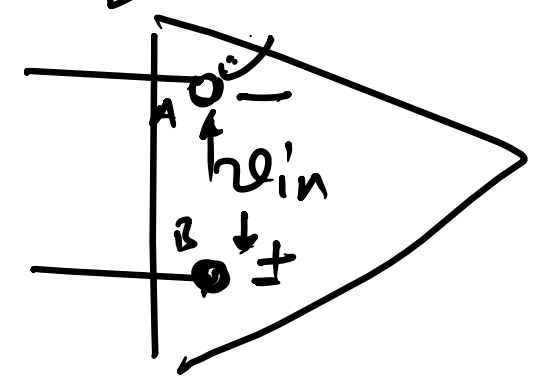
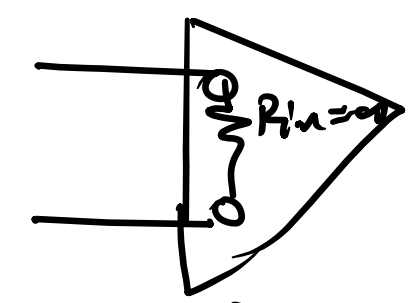
$$\Rightarrow \frac{v_i}{R_i} = v_A \left(\frac{1}{R_i} + \frac{1}{R_f} \right) - \frac{v_o}{R_f}$$

$\Rightarrow v_A = v_B \Rightarrow$ Virtual Short / Virtual gnd.

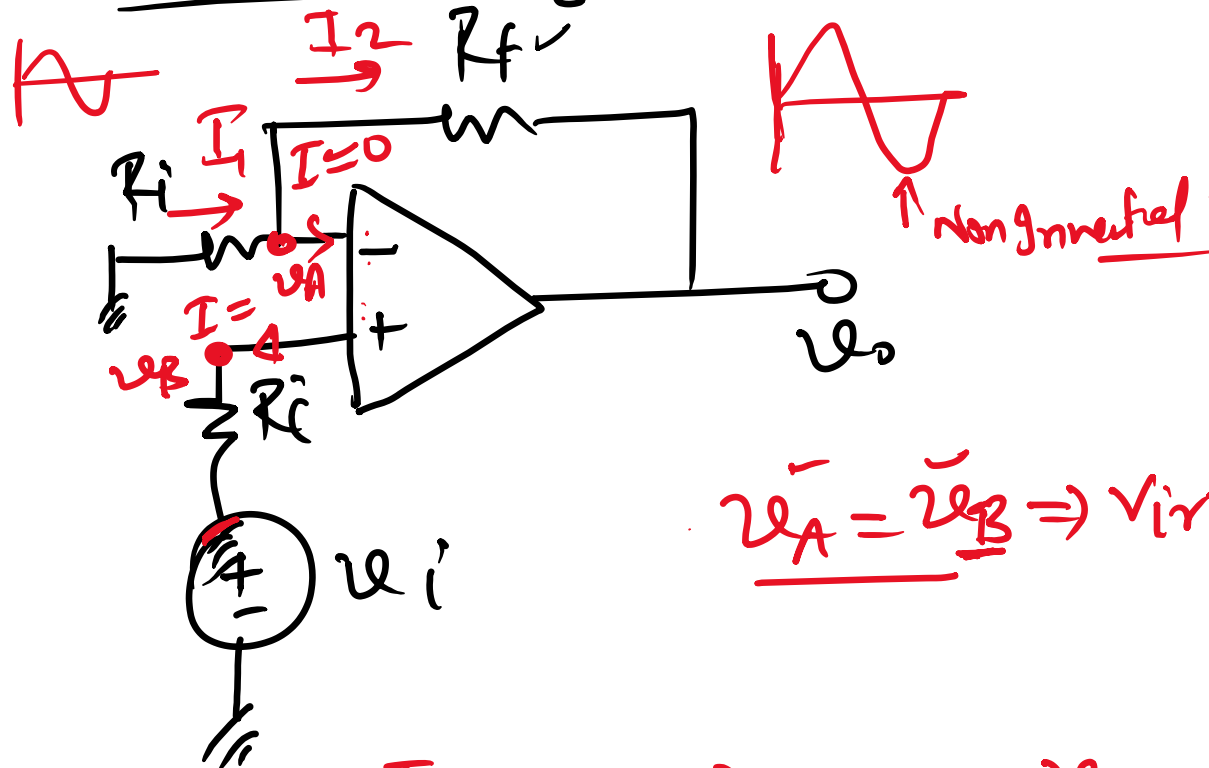
$$\underline{v_A = 0V}$$

$$\frac{v_i}{R_i} = -\frac{v_o}{R_f} \Rightarrow$$

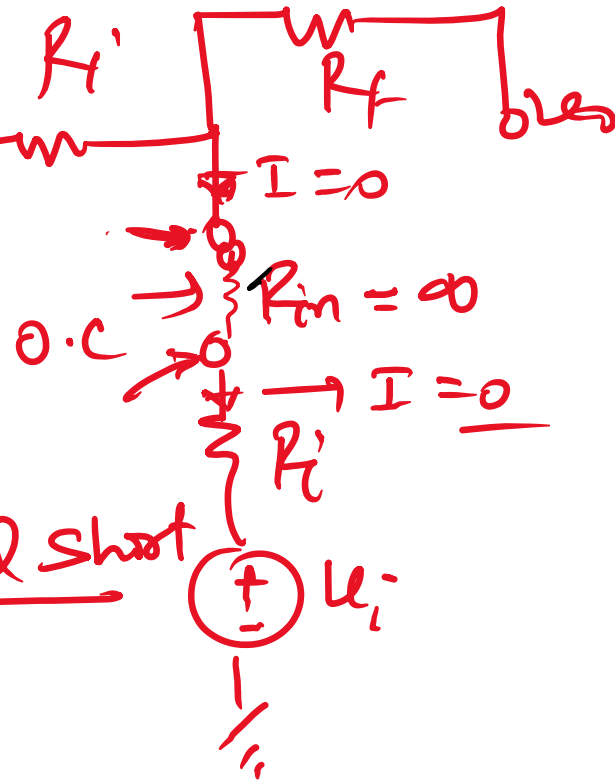
$$\underline{v_o = -\frac{R_f}{R_i} v_i}$$



Non-Inverting Amplifier



$\underline{u_A} = \underline{u_B} \Rightarrow$ Virtual short



$$\textcircled{A} \rightarrow I_1 = I_2 \Rightarrow \frac{0 - u_A}{R_i} = \frac{u_A - u_o}{R_f} \Rightarrow \frac{0 - u_i}{R_i} = \frac{u_i - u_o}{R_f}$$

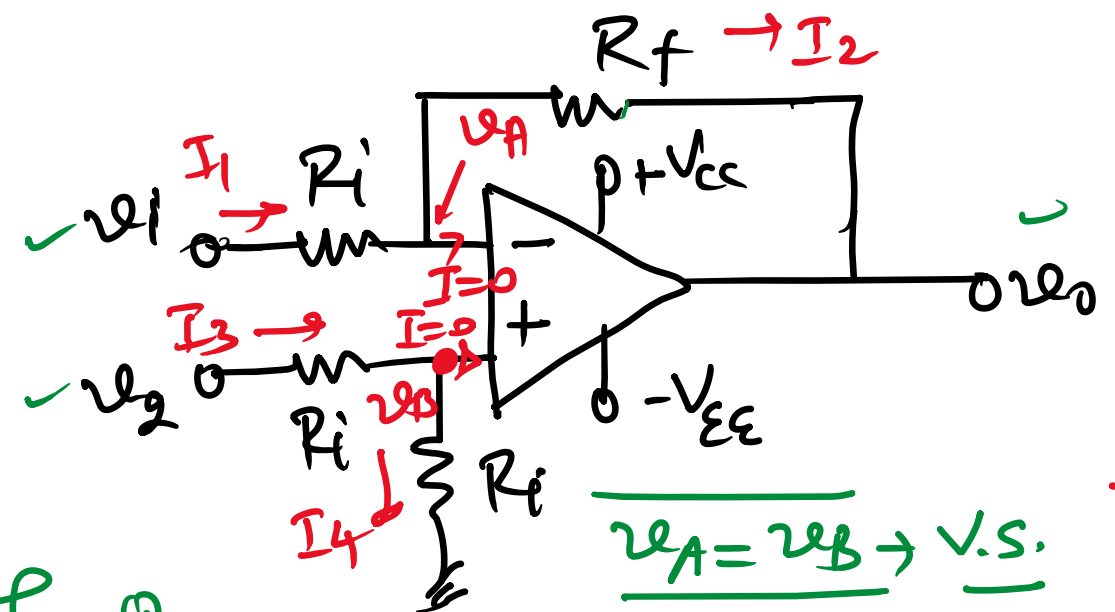
$$\Rightarrow \boxed{+u_o = +u_i \left(1 + \frac{R_f}{R_i} \right)}$$

$$\Rightarrow \boxed{A_v = \left[1 + \frac{R_f}{R_i} \right]}$$

Non-Inverting

- ① $R_i = \infty$
 - ② Virtual Short
- Operation of OP-Amp is based on these two chgs.

① Difference Amplifier - (Subtractor)



At node (A) -

$$I_1 = I_2 \Rightarrow \frac{v_1 - v_A}{R_i} = \frac{v_A - v_o}{R_f} \quad \text{--- (1)}$$

At node (B) -

$$I_3 = I_4$$

$$\frac{v_2 - v_B}{R_i} = \frac{v_B - 0}{R_i} \quad \text{--- (2)}$$

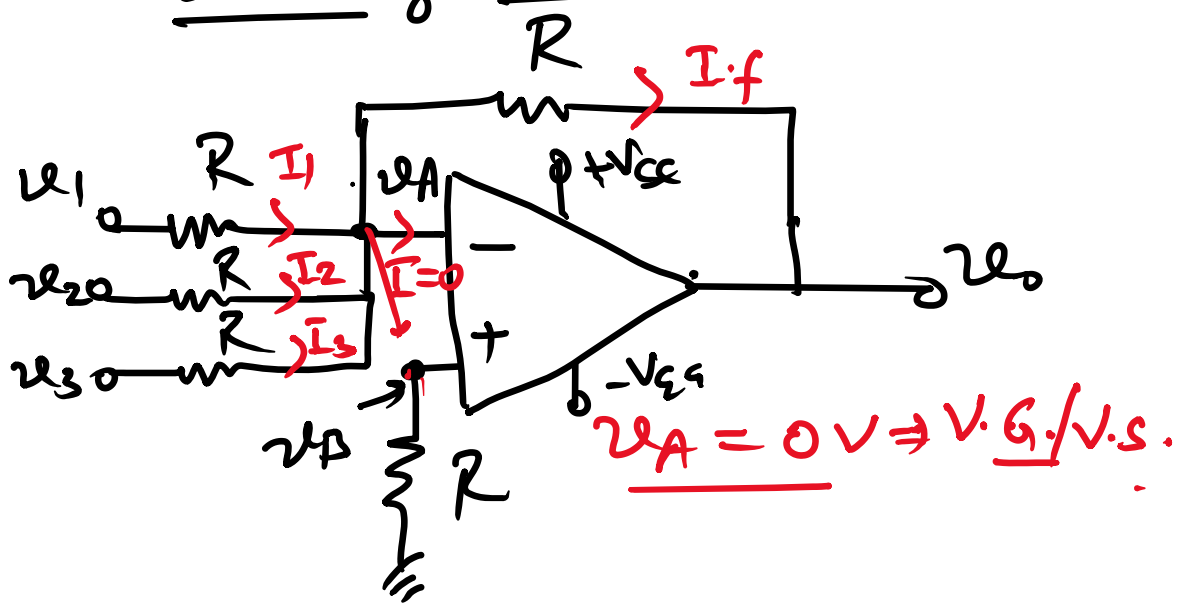
$$v_B = \frac{v_2}{2} \quad \text{--- (3)}$$

From (1) -

$$\frac{v_1 - v_2/2}{R_i} = \frac{v_2/2 - v_o}{R_f} \rightarrow \text{if } \underline{R_i = R_f} \Rightarrow \underline{v_1 - v_2 = -v_o} \Rightarrow \underline{v_o = v_2 - v_1}$$

$$\underline{v_1 - v_2 = -v_o} \Rightarrow \underline{v_o = v_2 - v_1}$$

Inverting adder -



At node - (A) \rightarrow

Adder \rightarrow Summed

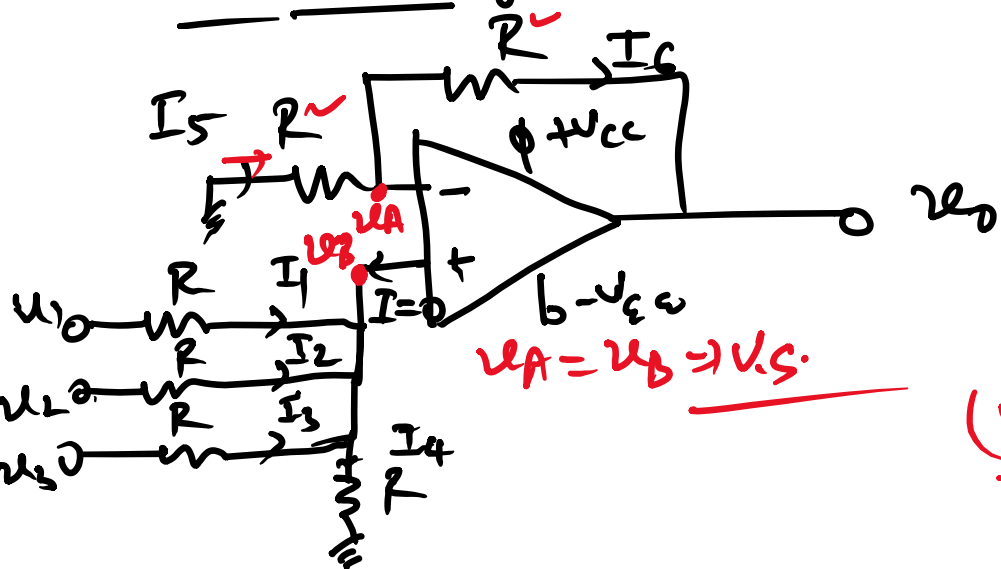
$$I_1 + I_2 + I_3 = I_4$$

$$\frac{v_1 - v_A}{R} + \frac{v_2 - v_A}{R} + \frac{v_3 - v_A}{R} = \frac{v_A - v_0}{R}$$

$$v_1 + v_2 + v_3 = -v_0$$

$$\boxed{v_0 = -(v_1 + v_2 + v_3)}$$

Non-Inverting adder -

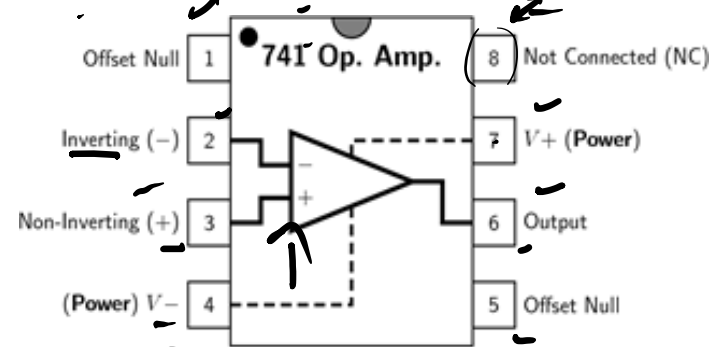
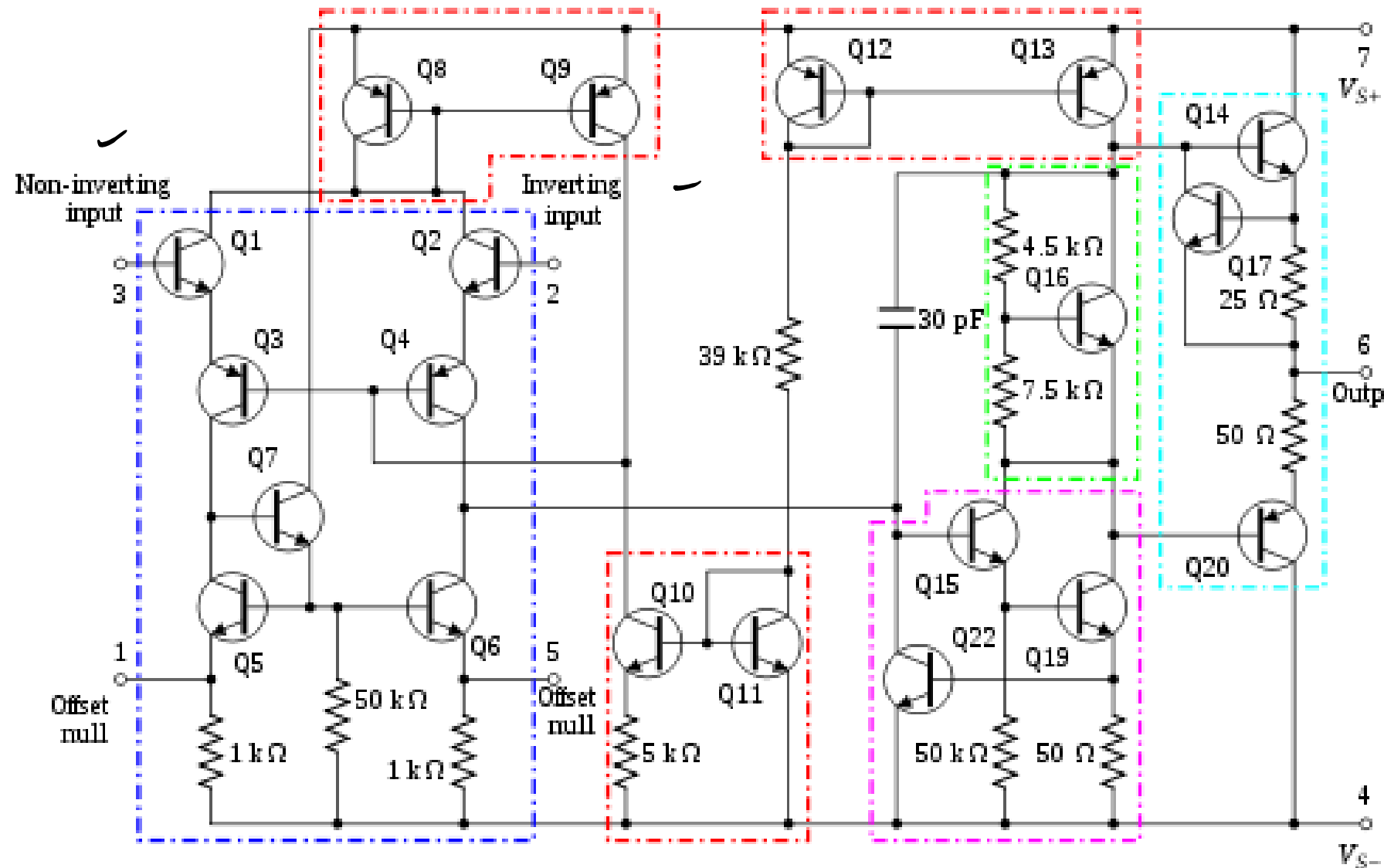
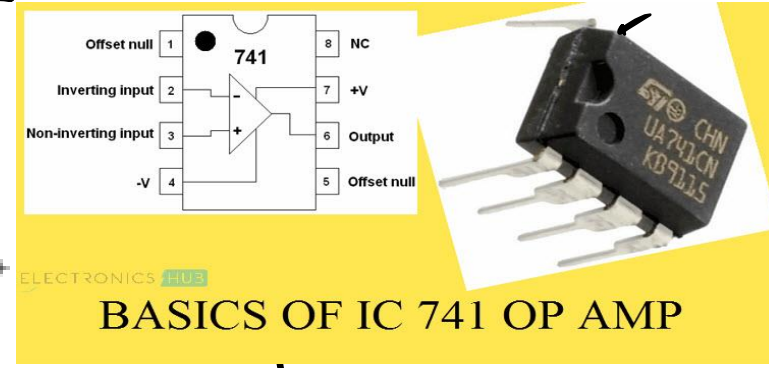


(A) $\rightarrow 0 - \frac{v_A}{R} = \frac{v_A - v_0}{R} \Rightarrow \frac{v_0}{2} = v_A$ (A)

(B) $\rightarrow I_1 + I_2 + I_3 = I_4 \Rightarrow \frac{v_1 - v_A}{R} + \frac{v_2 - v_A}{R} + \frac{v_3 - v_A}{R} = \frac{v_A - 0}{R}$ (B)

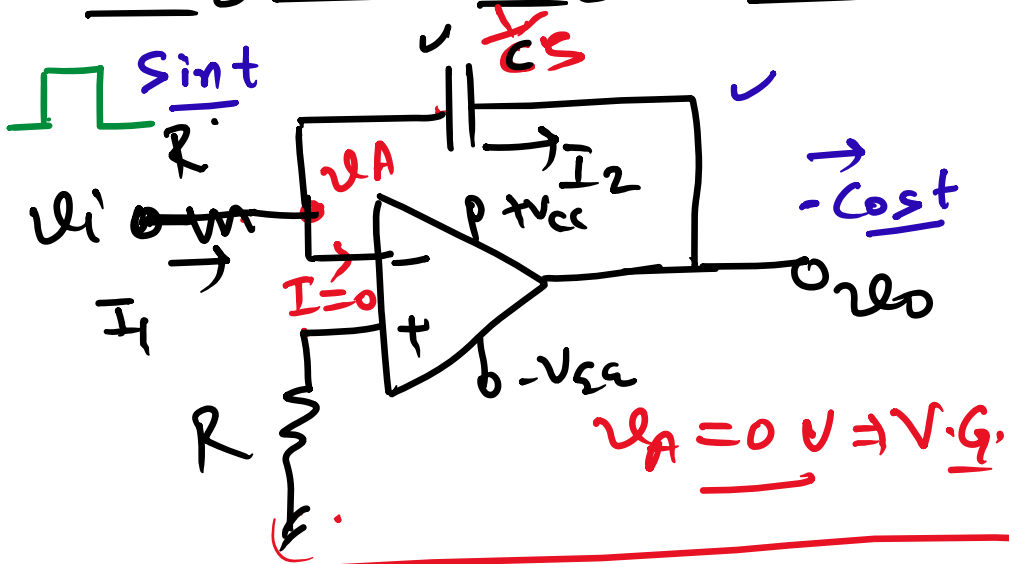
$$\frac{(v_1 + v_2 + v_3) - \frac{3v_0}{2}}{2} = \frac{v_0}{2} \Rightarrow \boxed{v_0 = \left(\frac{1}{2}\right)(v_1 + v_2 + v_3)}$$

Internal ckt diagram of OP-Amp (IC-741)



8-Pin Configuration

Integrator using OP-Amp



At node - (A) - ①

$$I_1 = I_2$$

$$\frac{u_i - 0}{R_i} = \frac{0 - u_o}{C s}$$

$$u_o C s = -\frac{u_i}{R_i}$$

$$u_o = -\frac{1}{RC(s)} u_i$$

$$u_o = -\frac{1}{RC} \int u_i dt$$

$$u_o = -\frac{1}{RC} \int_0^t 1 \cdot dt$$

$$u_o = -\frac{1}{RC} t$$

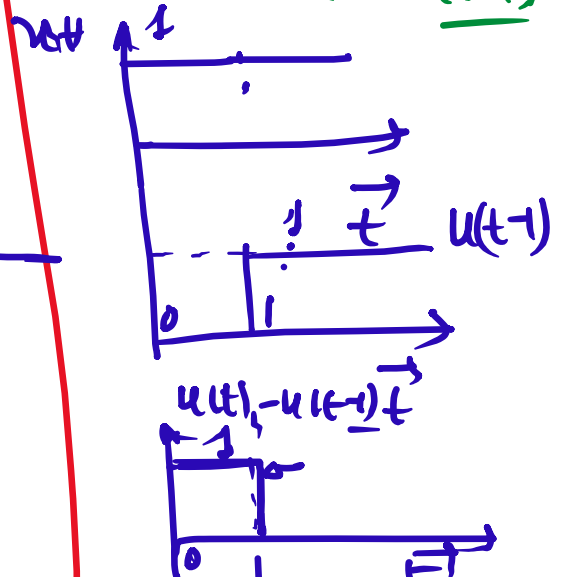
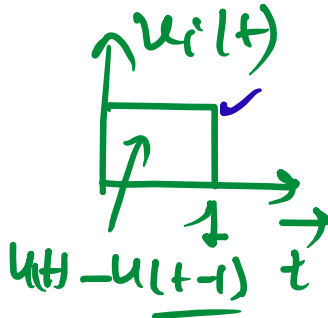
$$\rightarrow \frac{1}{s} \rightarrow \int dt$$

$$\rightarrow s' \rightarrow \frac{d}{dt}$$

Q

$$u_i = u(t) - u(t-1)$$

$$u_o = ?$$



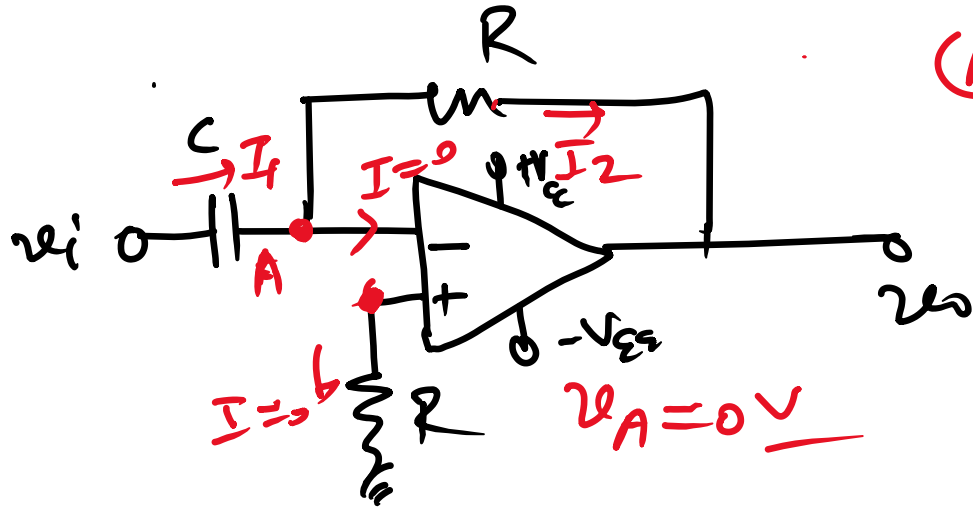
②

$$\frac{u_i - 0}{R} = C \frac{d(0 - u_o)}{dt}$$

$$\frac{u_i}{R} = -C \frac{du_o}{dt}$$

$$\Rightarrow u_o = -\frac{1}{RC} \int u_i dt$$

Differentiator - (capacitor in down position)



(A) \Rightarrow

$$I_1 = I_2 \Rightarrow C \frac{d(v_i - 0)}{dt} = \frac{0 - v_o}{R}$$

\Rightarrow

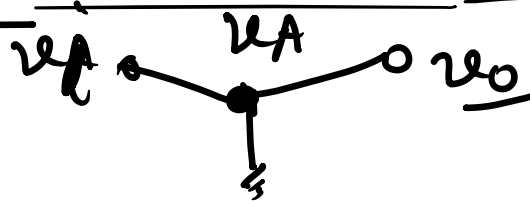
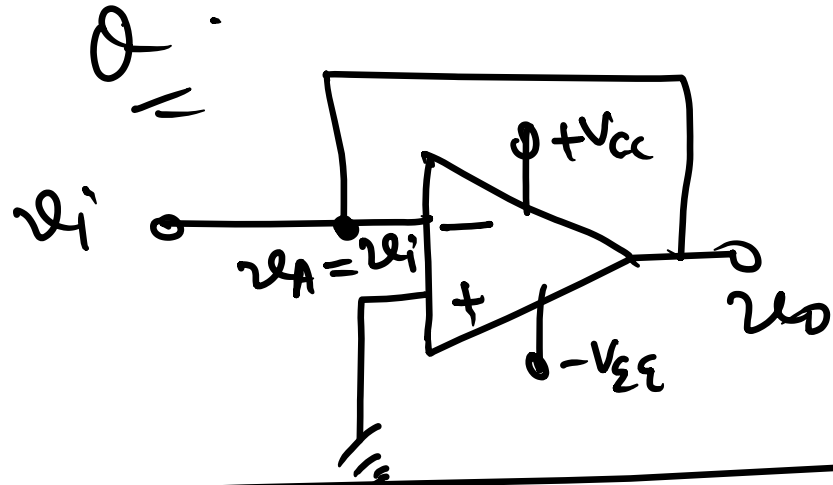
$$v_o = -RC \frac{dv_i}{dt}$$

Q. $v_i = \sin t$, calculate v_o for a differentiator ckt.

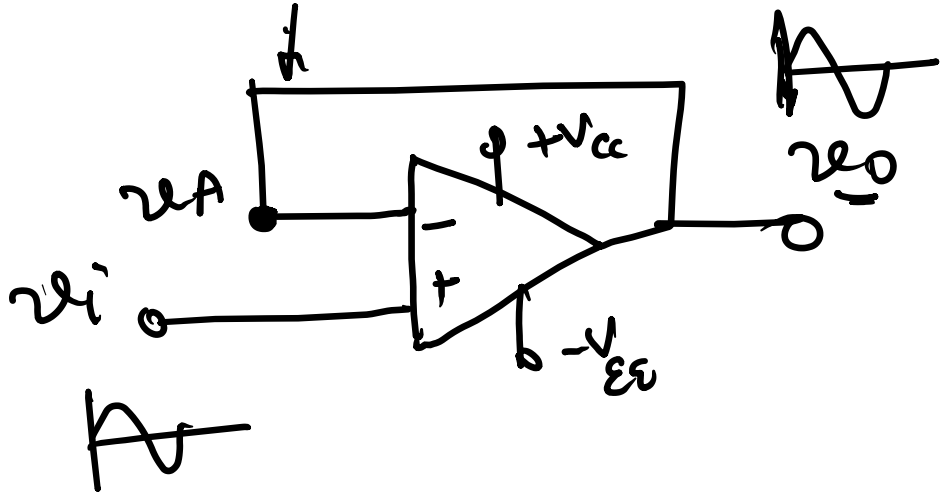
$$v_o = -RC \cos t$$

Q. $v_i = u(t) - u(t-1)$ \Rightarrow $v_o = ?$

Voltage follower -



$$\underline{\underline{v_A = ?}}$$



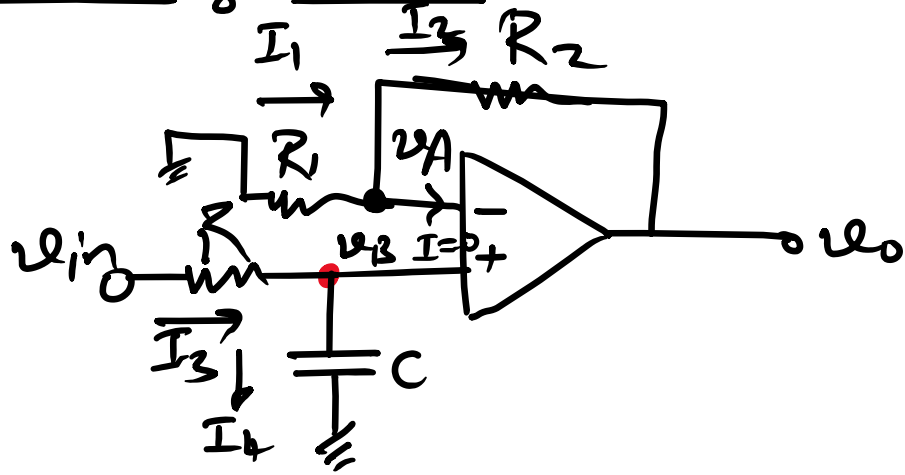
$$v_A = v_i$$

$$\boxed{v_o = v_i}$$

o/p voltage = I/p voltage
voltage follower ct

Active filters using OP-Amp -

- ① LPP
- ② HPF
- ③ BPF
- ④ BSF



take Laplace transform -

$$\frac{v_{in}}{R} - \frac{v_o R_1}{R(R_1 + R_2)} = sC \frac{v_o R_1}{R_1 + R_2}$$

$$\frac{v_{in}}{R} = \frac{v_o R_1}{R_1 + R_2} \left(\frac{1}{R} + sC \right)$$

$$= \frac{v_o (R_1)}{(R_1 + R_2)} \left(\frac{1 + sRC}{R} \right)$$

If $R_1 \gg R_2 \Rightarrow$

At node (A) -

$$I_1 = I_2$$

$$0 - \frac{v_A}{R_1} = \frac{v_A - v_o}{R_2}$$

$$v_A = \frac{v_o (R_1 R_2)}{R_2 (R_1 + R_2)} \Rightarrow 0$$

At node (B) -

$$I_3 = I_4$$

$$\frac{v_{in} - v_B}{R} = C \frac{d(v_B - 0)}{dt}$$

from - virtual short - $v_A = v_B$

$$\frac{v_{in}}{R} - \frac{v_o R_1}{R_1 + R_2} = \frac{C}{dt} \left(\frac{v_o R_1}{R_1 + R_2} \right)$$

$$\frac{v_{in}}{R} \neq v_o \left(\frac{1+sRC}{R} \right)$$

$$\boxed{v_o = \frac{v_{in}}{1+sRC}}$$

→ Transfer function -

$$\frac{v_o(s)}{v_{in}(s)} = \frac{1}{1+sRC}$$

$$\text{Gain} \Rightarrow \left| \frac{v_o(s)}{v_{in}(s)} \right| = \left| \frac{1}{1+sRC} \right|$$

$$s = j\omega \Rightarrow$$

$$\text{Gain} = \left| \frac{1}{1+j\omega RC} \right|$$

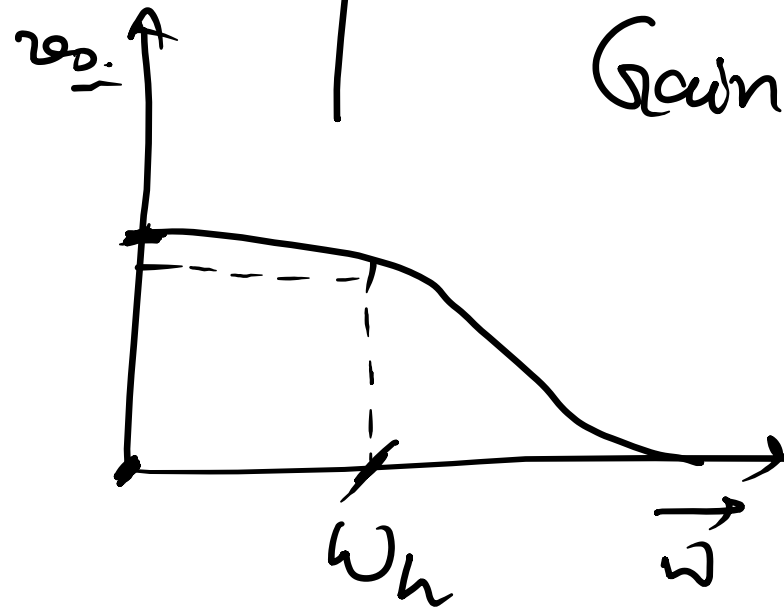
$$\boxed{\text{Gain} = \frac{1}{\sqrt{1+(\omega RC)^2}}}$$

① for $s = j\omega = 0$

$$v_o = v_{in}$$

② for $s = j\omega = \infty$

$$v_o = 0$$



Phase -

$$TF = \frac{1}{(1 + j\omega RC)}$$

$$\phi = \tan^{-1}(1) - \tan^{-1}\left(\frac{\omega RC}{1}\right)$$

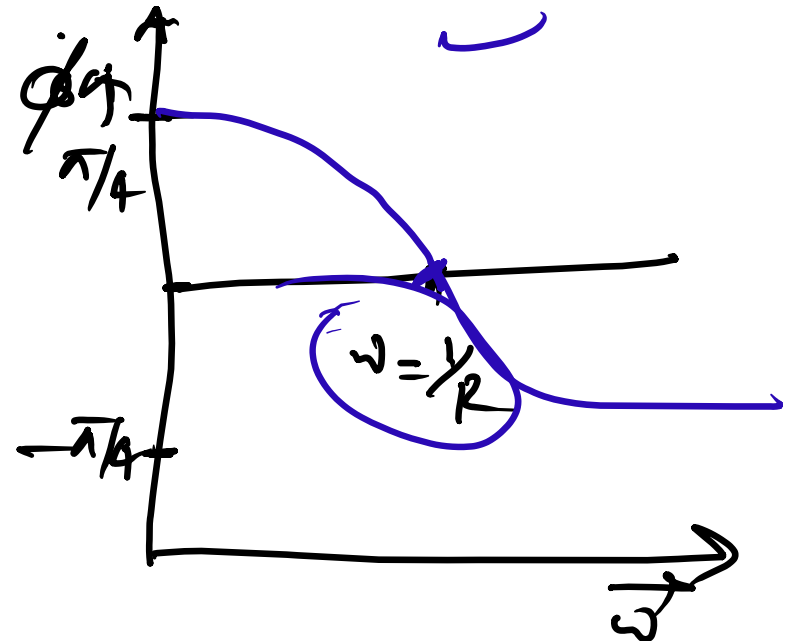
$$\phi = \frac{\pi}{4} - \tan^{-1}(\omega RC)$$

$$\phi = \frac{\pi}{4} - \tan^{-1}(\omega RC)$$

① $\omega = 0 \Rightarrow \phi = \frac{\pi}{4}$

② $\omega = 1/RC \Rightarrow \phi = \frac{\pi}{4} - \frac{\pi}{4} = 0$

③ $\omega = \infty \Rightarrow \phi = \frac{\pi}{4} - \frac{\pi}{2} = -\frac{\pi}{4}$



magnitude - $\text{Gain} = \frac{1}{\sqrt{1 + (\omega RC)^2}} \rightarrow \textcircled{1} \quad 3 \text{ dB } (10)$

Phase - $\phi = \frac{\pi}{4} - \tan^{-1}(\omega RC) \rightarrow \textcircled{2}$

Gain -

① $\omega = 0 \Rightarrow \text{Gain} = 1$

② $\omega = \frac{1}{RC} \Rightarrow \text{Gain} = \frac{1}{\sqrt{2}}$

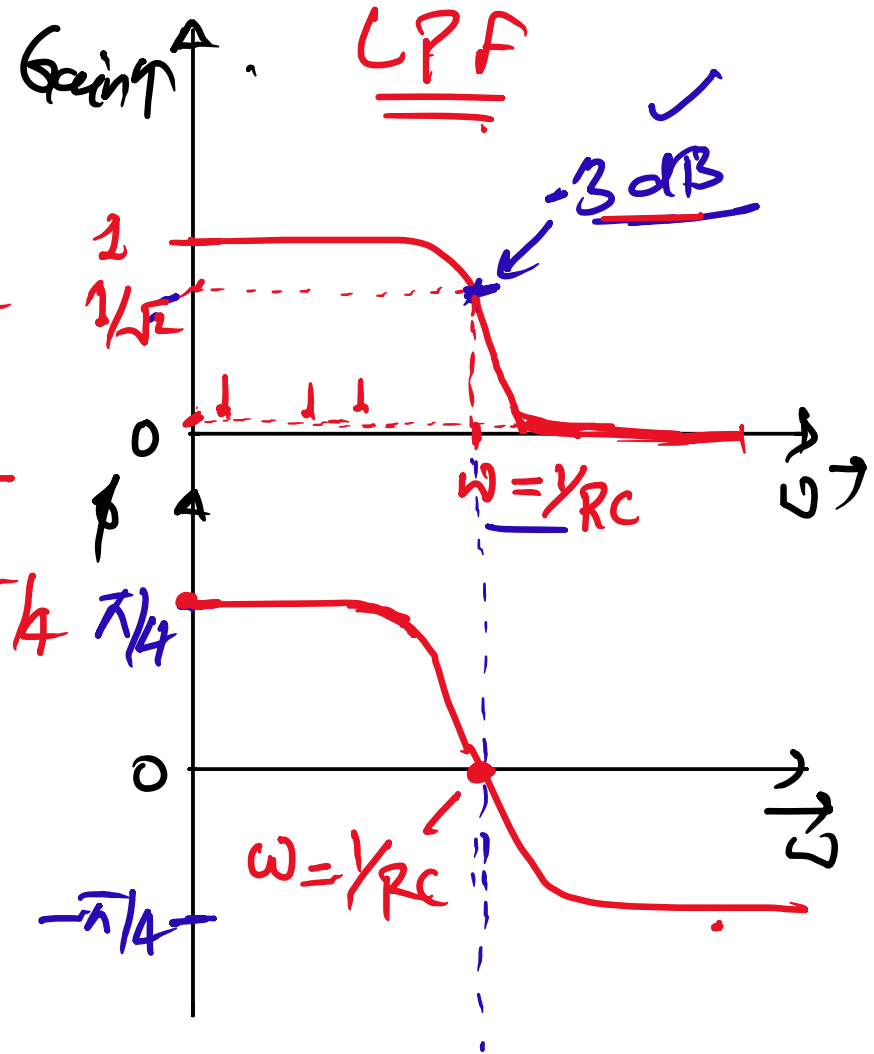
③ $\omega = \infty \Rightarrow \text{Gain} = 0$

Phase -

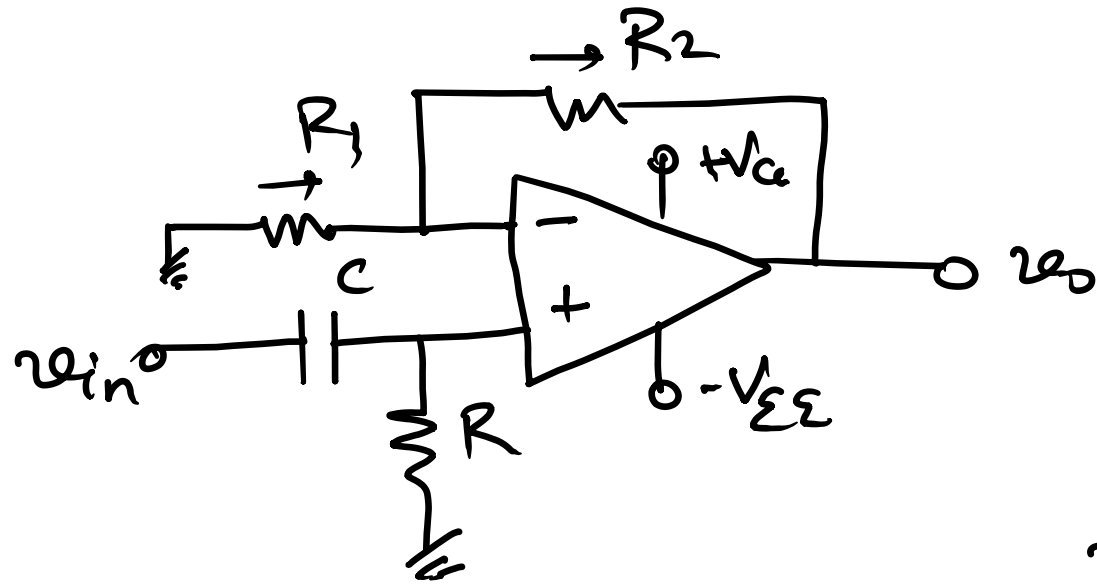
① $\omega = 0 \Rightarrow \phi = \frac{\pi}{4}$

② $\omega = \frac{1}{RC} \Rightarrow \phi = 0$

③ $\omega = \infty \Rightarrow \phi = -\frac{\pi}{4}$



High Pass Filter



Find ① Transfer function

$$TF = \frac{v_o(s)}{v_i(s)}$$

② Gain = $|TF| = \text{mag}(TF)$

③ Phase $\rightarrow \phi \Rightarrow$

Draw the plot for Gain & Phase.